# High Power RF MEMS Switch Technology

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HERIOT

The project partners include

1. Heriot-Watt University, U.K.

- 2. Rutherford Appleton Laboratory, U.K.
- 3. BAE Systems, U.K. BAE SYSTEMS



**Rutherford Appleton Laboratory** 



# **Outline**

- Introduction Motivation
- Concept for the development
- Design and Modelling
- Fabrication Processes
- Experiments
- Summary



# Why an interest in RF MEMS switches?

Many applications for RF MEMS switches

Radar systems for defence applications (5-94 GHz)
Automotive radars (24, 60 & 77 GHz)
Satellite communication systems (12-35 GHz)
Wireless communication Systems (0.8-6 GHz)
Instrument systems (0.01-50 GHz)



# **Advantages of RF MEMS switches**

- Near-zero power consumption
- High isolation
- Low insertion loss
- Low intermodulation products
- Low cost manufacture



### **Disadvantages**

- Relatively low stitching speed
- Low power handling
- High activation voltage
- Lack of suitable packaging and low reliability



# **Types of RF MEMS switch**

- Cantilever beam
- Membrane (Bridge)
- Others (such as thermal)



### **Cantilever beam**

- Silicon nitride beam
- Separate DC and RF electrodes
- RF electrode completes signal line





# **Membrane Switch**

- DC electrode in signal line
- Metallic membrane
- Appling DC voltage RF line shorted to ground i.e. high isolation







# MEMS Switch, Bridge Structure on CPW Fabricated at RAL







### MEMS Switch, Cantilever Structure on microstrip from BAE SYSTEMS





# MEMS switch, Bridge Structure on microstrip from BAE SYSTEMS





# Why this Project ?

- The above shown RF MEMS switches can only handle a low RF power, say up to a few hundreds milliwatts.
- High power RF MEMS switches, however, are of interest for many applications.
- It is the purpose of this collaborated project to design and demonstrate RF MEMS switches capable of handling 10 Watts RF power



# **Switch Specification**

Characteristic	Capability
Operating Frequency	Centred at 10 GHz with 10% bandwidth
Power Handling	10 W
Insertion Loss (on)	<0.2 dB
Return Loss (on)	>20 dB
Isolation (off)	>20 dB
Power Consumption	$<100 \ \mu W$
Threshold Voltage	<50 V
Switching Speed	10 µs
Characteristic Impedance	50 Ω



It is anticipated that there are great challenges in developing such high power RF MEMS switches.

How can we do it?



# **Switch Matrix Concept for the Development**



This approach allows configuring the switch matrix to be tailored for different power and isolation requirements while maintaining a low RF power level and low actuation voltage at each individual switching element.



#### 2 X 2 High Power Switch Matrix for a SPST Switch----Bridge Structure



Two pairs of bridge beams for a good stress balance





bias circuit, is 1.05 mm by 1.75 mm on a 250 $\mu$ m thick high resistivity (10K  $\Omega$ -cm), silicon wafer.





The switch is designed for operation at X-band. When all the switching elements are pulled down, the switch is in the RF "On" state. When all the beams are in the unactuated position, which is 3  $\mu$ m away from the RF transmission line on the wafer, the switch is in the RF "Off" state.







#### Current distribution at On-state for a 10 W input at 10 GHz





#### Charge distribution at Off-state for a 10 W input at 10 GHz





#### **Electromechanical Modelling**

□ Electromechanical modelling is important for the development of high power, RF MEMS switches.

□For our designs, it is important to characterise the single switching element, especially the RF contact.

□To this end, a simple mode is used first.



Points A and B are the two particular RF contact points on the beam, and d1 denotes the 'Off state' separation between the beam and the RF transmission line on the wafer.



# Electromechanical Modelling (simple model @ d1=3um)



•When V = 7.7 V, the beam or the contact B makes the first RF contact.

•The contact A only makes a contact until the actuation voltage is increased to 22 V, and this corresponds to the case in which the largest RF contact area is obtained.

•Further increasing the actuation voltage can results in an undesired deformation that causes the tip of the beam deformed upwards as shown for V = 25 V, and thus reduce the RF contact area.



# Electromechanical Modelling (simple model)

Displacement of the beam tip against the actuation voltage for d1 = 3, 4 and 5 um respectively



Each of the curves shows three distinct regions.

➤The first region shows that displacement increases against the actuation voltage before the beam makes the first RF contact.

The second region is a straight line indicating that the beam tip is in contact with the RF transmission line so that there is no change in the displacement.

➤The third region shows a decrease of the displacement, implying the beam tip has lost the contact with the RF line.

➤ The actuation voltage at the transition point between the first and second regions is the minimum voltage required for making an RF contact, which increases against the initial height of the beam.



# Electromechanical Modelling (simple model)

□For a given cantilever switch design and from the point of view of the RF contact, the desired actuation voltage would be that resulting in a largest contact area as well as a largest contact force.

The contact force for the proposed model can be estimated from the reaction force of the beam, which is shown in the next slide.



# Electromechanical Modelling (simple model)

Reaction force of the beam against the actuation voltage for the beam height = 3, 4 and 5 um respectively



In general, the reaction forces increase against the actuation voltage/beam displacement.

➤There is a clear transition point for each of the three curves shown, which correspond to the point at which the beam tip makes the first contact with the RF line.

>Up to the transition point the reaction force is only the spring or mechanical restoring force of the beam.

>On making first contact, a contact force is generated and increased as the actuation voltage increases.

➤In practise the contact force is expected to be smaller than the reaction force since the later includes the restoring force of the beam. In this case, the contact force is in a range of a few to several-tens of micro-Newtons.



# Electromechanical Modelling (Bridge model)





#### Join two cantilevers head-tohead to from a bridge model



#### Electromechanical Modelling (Bridge model @ d1=4um)





#### Actuation Voltage = 0 V

Actuation Voltage = 36 V



Actuation Voltage = 63 V

#### buckling



#### Electromechanical Modelling (Bridge model @ d1=4um)





#### Fabrication Flow Chart: Step 1-3





Fabrication Flow Chart: Step 4-5





#### Fabrication Flow Chart: Step 6-10

6. The 2<sup>nd</sup> lift-off

7. Sacrificial layer deposition, followed by contact metal Au define. ( Mask 3)

**8.** Thin Al (Alumnium) layer deposition.

**9**. gold layer deposited on the backside of wafer.

10.Post hole definition and sacrificial layer etched to open the post hole.(Mask 4)





#### Fabrication Flow Chart: Step 11-12

11. Photoresist and Allayer removed,followed by a 0.5umSiN deposition.

12. Photoresist patterning, Mask 5.





#### Fabrication Flow Chart: Step 13-15

**13**. SiN layer dry etch by RIE.

14. Resist removing, thenAu deposition. (~1um).(Mask 6)

**15**. Another layer of SiN (~0.5um) deposition and define.**Mask** 7 (layer 5)





#### Fabrication Flow Chart: Step 16

**16**. Sacrificial layer removing for released the cantilever.





#### Optical images of layer 1 and 2 on wafer





#### Optical images -- Before removing the sacrificial layer.

#### (a) Single switch.



#### (b) Double switch.





#### **Optical images**

#### (a) During plasma ashing



#### (b) After plasma ashing





The primary fabrication results reveal some problems, including high residual stress in switch beam and large variation of the beam height. The causes may lie in

(a) Material: SiN and Au do not match to each other

(b) Process: SiN deposition condition and the ashing process control (time, temperature etc.)

To overcome the problems, further investigation has been carried out into surface planarization on the sacrificial layer and low stress deposition with better-controlled gas flow, pressure and temperature.









Showing die mount assembly and DC and RF connections











Switch Die mounted on holder with DC and RF connections



DC test setup





#### DC actuating a test switch (video)







#### DC actuating a test switch array (video)







#### Switching time measurements



#### ~ 11 us for switching on

#### OFF time measurement



#### ~ 30 us for switching off





#### Diagram of microwave high power measurement set-up



The microwave power was generated by a synthesised source operating at x-band. The CW signal from the source passed through a variable 0-20 dB attenuator the output signal was then amplified by a Varian travelling-wave tube (TWT) amplifier. The output signal was from the TWT was fed into a 20 dB directional coupler which drew off a signal on the attenuated port for the input power measurement. The through port was coupled to the device under test (DUT) through a coax to waveguide transition. After the DUT end-to-end coaxial to waveguide transitions were used to provide DC voltage blocking to the power sensors for the through power measurement.





#### OFF state power measurements



The input power was increased until 35 dBm (about 3 W) without any noticeable change in the isolation of the switch across this range. The isolation remained at approximately -30 dB although this became lower with time due to slight drift in the gain of the TWT.





#### ON state power measurements

#### (a) Power-out vs. Power-in

#### (b) Insertion Loss



After removal of the calibration measurement, the switch insertion loss was measured to be in the range -1 dB to -0.5 dB. The drift in the TWT amplifier gain resulted in the lowering loss trend recorded. The device under test cold switched a power of 2.3W without damage occurring. In this instance the switch has not been tested to its maximum limit which remains to be investigated later.



### **Experiments**

The Si/SiO<sub>2</sub> combination was found to add significant loss to the S21 measurements before calibration, although both high resistive  $10K\Omega/cm$  Si and SiO<sub>2</sub> possess low loss.

On further investigation it was discovered that when these materials are combined the  $Si/SiO_2$  junction can produce an attenuation mechanism at high frequencies.

In order to address this problem a suitable alternative dielectric to  $Si/SiO_2$  will be required to provide adequate DC isolation. In recent literature the dielectric material BCB (BenzoCycloButene) has been successfully used in the fabrication for low loss high frequency applications.



# **Summary**

- A switch matrix concept has been introduced for the development of high power RF MEMS switches.
- An implementation of a high power, high isolation SPST switch with a 2X2 matrix has been demonstrated.
- Electromagnetic, electromechanical modelling as well as fabrication processes have been described.
- A number of interesting challenges remain in particular within the manufacturing process. The reduction of material stresses and improvement of yield as well thermal issue shall be addressed before this type of high power switch can be successfully developed.

