## Timing Diagrams

- Graphical representation of circuit behavior overtime
$\rightarrow$ illustrates the logic behavior of signals in a digital circuit as a function of time
- May be used as a device specification
$\rightarrow$ illustrates device performance
- May be used as a module or system specification
$\rightarrow$ identifies a requirement for system performance
- May be used as a tool in system analysis


Caution: terminology may vary slightly between vendors.
Set-up time , $\mathrm{t}_{\mathrm{su}}$ : the minimum length of time that a signal must be valid at a circuit input before a second triggering signal arrives at a second input.

Usually a clock

Delay time , $\mathrm{t}_{\mathrm{co}}$ : the length of time that a circuit requires for its output(s) to begin to change in response to a triggering signal arriving at an input. (also called propagation delay)

Hold time , $\mathrm{t}_{\mathrm{no}}$ : the minimum length of time that a signal must be kept valid at a circuit input after a triggering signal has been received at a second input.
Timing skew, $\mathrm{t}_{\text {skew }}$ : the maximum range of times over which a particular signal transition can occur.
-- Due to variations in driver output impedance
-- Problems in clock distribution

## Nomenclature for Timing Diagrams

| Symbol | Input | Output |
| :---: | :---: | :---: |
|  | The input must be valid | The output will be valid |
|  | If the input were to fall | Then the output will fall |
| $\underline{\square}$ | If the input were to rise | Then the output will rise |
| P0x | Don't care. it will work regardless | Don't know, the output value is indeterminate |
| - | Nonsense | High impedence, tristate. HiZ. <br> Not driven, floating |

## Example Timing Diagram

- Functional Timing Diagram (idealized)
$\rightarrow$ assumes zero delays
$\rightarrow$ simply demonstrates logic relations


Arrows show cause and effect which input transactions cause which output transactions,
especially in complex timing diagram

## Example Timing Diagram

- Timing Diagram (more realistic)
$\rightarrow$ shows delays using typical or maximum values

- $\mathrm{t}_{\mathrm{pHL}}=$ HIGH to LOW propagation delay
- $t_{\text {pLH }}=$ LOW to HIGH propagation delay


## 74LS244 Buffer 1



| PARAMETER | TEST CONDITIONS |  | 'LS240 |  |  | 'LS241, 'LS244 |  |  | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | MIN | TYP | MAX | MIN | TYP | MAX |  |
| tpLH | $\mathrm{R}_{\mathrm{L}}=667 \Omega$, | $C_{L}=45 \mathrm{pF}$ |  | 9 | 14 |  | 12 | 18 | ns |
| $\mathrm{t}_{\text {PHL }}$ |  |  |  | 12 | 18 |  | 12 | 18 |  |

- Propagation delay $=12 n S$ (enabled) - typical value
$\rightarrow$ Analysis (design) should use MAXIMUMyalues for worst case.


## 74LS244 Buffer 2



- tri-state buffer (see later slide)
- Propagation delay $=12 n S$ (enabled)


| PARAMETER | TEST CONDITIONS |  | 'LS240 |  |  | 'LS241, 'LS244 |  |  | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | MIN | TYP | MAX | MIN | TYP | MAX |  |
| tPLH | $\mathrm{R}_{\mathrm{L}}=667 \Omega$, | $C_{L}=45 \mathrm{pF}$ |  | 9 | 14 |  | 12 | 18 | ns |
| $\mathrm{t}_{\text {PHL }}$ |  |  |  | 12 | 18 |  | 12 | 18 |  |
| tpZL | $\mathrm{R}_{\mathrm{L}}=667 \Omega$, | $C_{L}=45 \mathrm{pF}$ |  | 20 | 30 |  | 20 | 30 | ns |
| tPZH |  |  |  | 15 | 23 |  | 15 | 23 |  |
| tPLZ | $R_{L}=667 \Omega$, | $C_{L}=5 \mathrm{pF}$ |  | 10 | 20 |  | 10 | 20 | ns |
| tpHz |  |  |  | 15 | 25 |  | 15 | 25 |  |

## Karnaugh Map



- Karnaugh Maps (K-Maps) are a graphical method of visualizing the 0's and 1's of a boolean function
$\rightarrow$ K-Maps are very useful for performing Boolean minimization.
- Will work on 2, 3, and 4 variable K-Maps in this class.
$\rightarrow$ Variable-Entered-Maps will be used for systems with more than 4 variables.
- Karnaugh maps are much easier to use than boolean equations for minimization.


## Karnaugh Map

- Karnaugh Map Method
$\rightarrow$ K-map is an alternative method of representing the truth table that helps visualize adjacent terms in up to 6 dimensions
 K-map

$\rightarrow$ Numbering Scheme: 00, 01, 11, 10
$\rightarrow$ Gray Code -- only a single bit changes one code word to the next



## Boolean Adjacency

- Note on the three variable map:


WRONG!!!

- Each square on the 3-variable map is Boolean Adjacent. Adjacent squares
- only differ by ONE BOOLEAN VARIABLE
- Although drawn as a 2-D diagram the edges wrap round left-right and top-bottom.



## Plotting 4-Variable Functions

| Plotting 4-Variable Functions |  |  |  |  |  |  |  | $\Psi_{0}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Row\|A B C D |  | F(A,B,C,D) | CD ${ }^{\text {AB }}$ |  | 01 |  |  |  |
| 0 | 0000 |  |  |  |  |  |  |  |
| 1 | 0 0001 | $?$ | 00 | ? |  | ? | ? | ? |  |
| 2 | $\begin{array}{lllll}0 & 0 & 1 & 1 \\ 0 & 0 & 1 & 1\end{array}$ | $?$ | 01 | ? | ? | ? | ? |  |
| 4 | 0100 | ? | 11 | ? | ? | ? | ? |  |
| 5 | 0101 | ? |  |  |  |  |  |  |
| 6 | 0110 | ? | 10 | ? | ? | ? | ? |  |
| 7 | 0111 | ? |  |  |  |  |  |  |
| 8 | 1000 | ? |  | $\mathrm{B}_{0}$ |  |  |  |  |
| 9 | 1001 | ? | CD |  | 01 | 11 | 10 |  |
| 10 | 1010 | ? | 00 | r0 | r4 | r12 | r8 |  |
| 11 | $\begin{array}{lllll}1 & 0 & 1 & 1 \\ 1 & 1 & 0 & 0\end{array}$ | ? | 01 | r1 | r5 | r13 | r9 |  |
| 12 | $\begin{array}{llll}1 & 1 & 0 & 0 \\ 11 & 1 & 0 & 1\end{array}$ | ? | 11 | r3 | r7 | r15 | r11 |  |
| 14 | 1110 | ? |  |  |  |  |  |  |
| 15 | 1111 | ? | 10 | r2 | r6 | r14 | r10 |  |
|  |  |  | 14 |  |  |  |  | Digtala Design |



| Row | A B C | F(A,B,C) | ${ }^{\text {A }}$ |  | 1 |
| :---: | :---: | :---: | :---: | :---: | :---: |
| 0 | 000 | 0 | ${ }^{\text {BC }}$ | 0 | 0 |
| 1 | $\begin{array}{llll}0 & 0 & 1 \\ 0 & 1 & 0\end{array}$ | 0 | 00 |  | 0 |
| 2 | 010 |  |  | 0 | 0 |
| 3 | 011 | 0 | 11 | 0 | 0 |
| 4 | 100 | 0 | 10 |  |  |
| 5 | 101 | 0 |  |  |  |
| 6 | 110 |  |  |  |  |
| 7 | 111 | 0 | F(A,B,C | $\begin{array}{r} C)=\overline{1} \\ =\bar{A} \\ =E \\ =E \end{array}$ | $\begin{aligned} & \Sigma_{1} \\ & \overline{\mathrm{~A}} \mathrm{~B} \\ & \mathrm{BC} \\ & \mathrm{BC} \end{aligned}$ |

Boolean adjacency can be used to minimize functions!

## Simplification using K-Maps



- Grouping blocks of '1'
$\rightarrow$ A group must consist of $16,8,4,2$ or 1 cells
$\rightarrow$ Each cell must be horizontally and/or vertically adjacent to cells in the other group
$\rightarrow$ Always include the largest number of '1's in a group
$\rightarrow$ Each '1' in the map should be included in a group
$\rightarrow$ Groups can overlap
$\rightarrow$ Map edge cells connect in a loop to cells at the opposite edge
- Naming groups
$\rightarrow$ The product description for a group will include ALL variables that are CONSTANT over the group
$\rightarrow$ For example, for a 4-variable map
> An 8 -cell group is described by a 1 -variable product term
> An 4 -cell group is described by a 2 -variable product term
> An 2-cell group is described by a 3 -variable product term
> An 1 -cell group is described by a 4 -variable product term


## Groupings on K-Maps

- Grouping can be read DIRECTLY as "BC" by looking at what is COMMON within the circled group.



## Example Groupings on 3-Variable K-Maps

| ${ }^{\text {A }}$ | 0 | 1 |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| BC |  |  | $\mathbf{F}(\mathbf{A}, \mathrm{B}, \mathrm{C})=$ |  |  | $\mathbf{F}(\mathbf{A}, \mathrm{B}, \mathrm{C})=$ |
| 00 | 1 | 0 |  |  |  |  |
| 01 | 1 | 0 |  |  |  |  |
| 11 | 0 | 0 | BC | 0 | 1 |  |
| 10 | 0 | 0 | 00 | 1 | 1 |  |
| 10 | 0 | 0 | 01 | 0 | 0 |  |
| ${ }^{\text {A }} 0$ |  |  | 11 | 0 | 0 |  |
|  |  |  | 10 | 1 | 1 |  |
| 00 | 1 | 1 | $\mathbf{F}(\mathbf{A}, \mathrm{B}, \mathrm{C})=$ |  |  |  |
| 01 | 1 | 1 |  |  |  |  |
| 11 | 0 | 0 |  |  |  |  |
| 10 | 0 | 0 |  |  |  |  |

Try to cover all ' 1 's with largest possible groupings.

| ${ }^{\text {A }}$ | 0 | 1 |
| :---: | :---: | :---: |
| 00 | 1 | 0 |
| 01 | 1 | 1 |
| 11 | 0 | 0 |
| 10 | 0 | 0 |

$F(A, B, C)=$

| ${ }^{\text {A }} 0$ |  |  |
| :---: | :---: | :---: |
| BC |  |  |
| 00 | 0 | 1 |
| 01 | 0 | 0 |
| 11 | 1 | 0 |
| 10 | 1 | 0 |

Groupings of only a single ' 1 ' are ok if larger groupings cannot be found.

$$
\mathbf{F}(\mathbf{A}, \mathbf{B}, \mathrm{C})=
$$

## Illegal Groupings



$$
\bar{A} \bar{B} \bar{C}+A \bar{B} C=\bar{B}(\bar{A} \bar{C}+A C)
$$

- Valid groupings will always be a power of 2.
$\rightarrow$ will cover $1,2,4,8$, etc minterms.

| $\mathrm{CD}^{\mathrm{AB}} \mathbf{0 0}$ |  | 01 | 11 | 10 |
| :---: | :---: | :---: | :---: | :---: |
| 00 | 0 | 0 | 0 | 0 |
| 01 | 0 | 0 | 0 | 0 |
| 11 | 1 | 0 | 1 | 0 |
| 10 | 1 | 1 | 0 | 1 |

$\mathbf{F}(\mathbf{A}, \mathbf{B}, \mathbf{C}, \mathbf{D})=$

| Other Groupings |  |  |  |  |  | $\xrightarrow{+0}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  |  |  | $\left.\right\|_{c}$ |  |
|  |  |  |  |  |  |  |
| $\text { D } \left\lvert\, \begin{aligned} & 00 \\ & 01 \\ & 11 \\ & 10 \end{aligned}\right.$ | 1 | 0 | 0 | 1 |  |  |
|  | 1 | 0 | 0 | 1 |  |  |
|  | 1 | 0 | 0 | 1 |  |  |
|  | 1 | 0 | 0 | 1 |  |  |
|  |  |  |  |  |  |  |
| $\mathbf{F}(\mathbf{A}, \mathbf{B}, \mathbf{C}, \mathbf{D})=$ |  |  |  |  |  |  |
| 23 |  |  |  |  |  | Digital Design |

More than one way to group.....

| $\begin{array}{ccccc}\mathrm{AB}^{\text {AB }} & \\ \text { CD } & & 11 & 10\end{array}$ |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  |  |  |  |
| 00 | 1 | 1 | 1 | 1 | $\mathbf{F}(\mathbf{A}, \mathbf{B}, \mathbf{C}, \mathbf{D})=\overline{\mathrm{B}} \mathrm{D}+\overline{\bar{C}} \bar{D}+C \bar{D}$ |
| 01 | T | 0 | 0 | 1 |  |
| 11 | 1 | 0 | 0 | 1 |  |
| 10 | 1 | 1 | 1 | 1. |  |



## Four Corner Grouping on 4-Variable Map

| AB |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: |
| CD 0000 |  |  |  |  |
| 00 | 1 | 0 | 0 | 1 |
| 01 | 0 | 0 | 0 | 0 |
| 11 | 0 | 0 | 0 | 0 |
| 10 | 1 | 0 | 0 | 1 |

4 Corner
grouping is valid
on four variable map

$$
\mathbf{F}(\mathbf{A}, \mathbf{B}, \mathbf{C}, \mathbf{D})=
$$

## K-map Definitions



- Implicant
$\rightarrow$ Any single 1 or any group of 1 's is called an implicant of $F$.
$\rightarrow$ Any possible grouping of ' 1 's is an implicant.

- Prime Implicant
$\rightarrow$ A implicant that cannot be combined with some other implicant to eliminate a variable


## Minimum Sum-Of-Products (SOP)

- The minimum SOP expression consists of some (but not necessarily all) of the prime implicants of a function.
- If a SOP expression contains a term which is NOT a prime implicant, then it CANNOT be minimum.


## Prime Implicants



EACH of these coverings is a PRIME
IMPLICANT (i.e. cannot be reduced)
$\bar{B} \bar{C}, \bar{A} \bar{C} D, \bar{A} \bar{B} D$

Minimum SOP will have some or all of these prime implicants.
The included prime implicants must cover all of the ONEs.
$\mathbf{F}(\mathbf{A}, \mathbf{B}, \mathbf{C}, \mathbf{D})=\mathrm{B} \bar{C}+\overline{\mathrm{A}} \bar{B} D \quad$ (minimum set of PIs)
$=B \bar{C}+\bar{A} \bar{B} D+\bar{A} \bar{C} D \quad$ (valid set of PIs, but not minimum)
$\neq \bar{A} \bar{B} D+\bar{A} \bar{C} D$
(both PI's, but all '1's not included!)

$$
\mathbf{F}(\mathbf{A}, \mathbf{B}, \mathbf{C}, \mathrm{D})=\mathrm{B} \bar{C}+\overline{\mathrm{A}} \overline{\mathrm{~B}} \mathrm{D} \quad \text { (minimum \# of PIs) }
$$

- NON-ESSENTIAL prime implicant
$\rightarrow$ Prime Implicant $\bar{A} \bar{C} D$ is non-essential because its ' 1 's are covered by other Pls. A PI is ESSENTIAL if it covers a MINTERM that cannot be covered by any other PI.

An example with more than one solution


Recall that a covering is a Prime Implicant if it cannot be combined with another covering to eliminate a variable.

## Two Solutions

## $\underset{\sim}{\stackrel{3}{\rightleftarrows}}$



EACH solution is equally valid.

$$
\mathbf{F}(\mathbf{A}, \mathbf{B}, \mathbf{C}, \mathbf{D})=\bar{A} \bar{C}+\mathrm{ACD}+\overline{\mathrm{A}} \mathrm{BD}
$$

Essential PIs
Non-Essential
PIs

$\mathbf{F}(\mathbf{A}, \mathbf{B}, \mathbf{C}, \mathbf{D})=\bar{A} \bar{C}+A C D+B C D$

## Minimal Solution

- A minimal SOP will consist of prime implicants.
- A minimal SOP equation will have all of the essential prime implicants on the map. By definition, these cover a minterm that may not be covered by some other prime implicant.
- The minimal SOP equation may or may not include nonessential prime implicants. It will include non-essential prime implicants if there are ' 1 's remaining that have not been covered by an essential prime implicant.

